# Intel® Atom ${ }^{\text {TM }}$ Processor $330^{\text {® }}$ Series <br> Datasheet 

For systems based on Entry Level Desktop Platform for '08
Revision 003
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## Contents

1 Introduction ..... 6
1.1 Intel ${ }^{\circledR}$ Atom ${ }^{\text {TM }}$ Processor 300 Series Features ..... 6
1.2 Terminology ..... 7
1.3 References ..... 8
2 Low Power Features ..... 9
2.1 Clock Control and Low-power States ..... 9
2.1.1 Thread Low-power State Descriptions ..... 10
2.1.1.1 Thread CO State ..... 10
2.1.1.2 Thread C1/AutoHALT Powerdown State ..... 10
2.1.1.3 Thread C1/MWAIT Powerdown State ..... 10
2.1.2 Package Low-power State Descriptions ..... 11
2.1.2.1 Normal State ..... 11
2.1.3 Front Side Bus ..... 11
3 Electrical Specifications ..... 12
3.1 FSB and GTLREF ..... 12
3.2 Power and Ground Pins ..... 12
3.3 Decoupling Guidelines ..... 13
3.3.1 VCCP Decoupling ..... 13
3.3.2 FSB AGTL+ Decoupling ..... 13
3.4 Voltage Identification and Power Sequencing ..... 14
3.5 Catastrophic Thermal Protection ..... 15
3.6 Reserved and Unused Pins ..... 15
3.7 FSB Frequency Select Signals (BSEL[2:0]) ..... 16
3.8 FSB Signal Groups ..... 16
3.9 CMOS Asynchronous Signals ..... 18
3.10 Maximum Ratings ..... 18
3.11 Processor DC Specifications ..... 19
3.12 AGTL+ FSB Specifications ..... 23
4 Package Mechanical Specifications and Ball Information ..... 24
4.1 Package Mechanical Specifications ..... 24
4.1.1 Package Mechanical Drawings ..... 25
4.1.2 Package Loading Specifications ..... 25
4.1.3 Processor Mass Specifications ..... 25
4.2 Processor Pinout Assignment ..... 25
4.3 Signal Description ..... 31
5 Thermal Specifications and Design Considerations ..... 39
5.1 Thermal Specifications ..... 39
5.1.1 Thermal Diode ..... 40
5.1.2 Intel ${ }^{\circledR}$ Thermal Monitor ..... 42
5.1.3 Digital Thermal Sensor ..... 44
5.1.4 Out of Specification Detection ..... 44
5.1.5 PROCHOT\# Signal Pin ..... 44
6 Debug Tools Specifications ..... 46

## Figures

2-1 Thread Low-power States ..... 9
4-2 Package Mechanical Drawing ..... 25
4-3 Pinout Diagram (Top View, Left Side) ..... 26
4-4 Pinout Diagram (Top View, Right Side) ..... 27
Tables
2-1 Coordination of Thread Low-power States at the Package Level ..... 10
3-2 Voltage Identification Definition ..... 14
3-3 Processor VID Pin to VRD11 VID Pin Mapping ..... 15
BSEL[2:0] Encoding for BCLK Frequency ..... 16
3-5 FSB Pin Groups ..... 17
3-6 Processor Absolute Maximum Ratings ..... 18
3-7 Voltage and Current Specifications for the processor. ..... 19
3-8 FSB Differential BCLK Specifications ..... 20
3-9 AGTL+/CMOS Signal Group DC Specifications ..... 21
3-10 Legacy CMOS Signal Group DC Specifications ..... 22
3-11 Open Drain Signal Group DC Specifications ..... 22
4-12 Pinout Arranged By Signal Name. ..... 28
4-13 Signal Description ..... 31
5-14
Thermal Diode Interface ..... 41
5-16 Thermal Diode Parameters using Transistor Model ..... 41

Revision History

| Revision Number | Description | Date |
| :---: | :---: | :---: |
| -001 | - Initial Release | September 2008 |
| -002 | - Update pin-map <br> - Add SSSE3 <br> - Changed A[35:2] to A[32:2] <br> - Changed Vboot <br> - Changed Ron and Rodt <br> - Removed L2 Dynamic Cache Sizing | February 2009 |
| -003 | - Updated Table 3-7: Removed information on dI/dt | April 2010 |

§

Introduction

## 1 I ntroduction

The Intel ${ }^{\circledR}$ Atom ${ }^{\text {TM }}$ processor 330 series is built on 45 -nanometer Hi-k process technology. This document contains electrical, mechanical and thermal specifications for the processor.

Note: In this document, the Intel Atom processor 330 series will be referred to as the processor. Intel chipsets shall be referred as GMCH and ICH respectively.

### 1.1 Intel ${ }^{\circledR}$ Atom ${ }^{\text {TM }}$ Processor 300 Series Features

- Available at 1.6 GHz ,
- New dual-core processor for Entry Level Desktop Platform PCs with enhanced performance
- On die, primary 32-kB instructions cache and $24-\mathrm{kB}$ write-back data cache, per core
- 533-MHz Source-Synchronous front side bus (FSB)
- Threading enabled
- On die 512-kB, 8-way L2 cache, per core: total of 1 MB cache
- Support for IA-32 and Intel ${ }^{\circledR} 64$ architecture
- Streaming SIMD Extensions 2 and 3 (SSE2 and SSE3) and Supplemental Streaming SIMD Extensions 3 (SSSE3) support
- Micro-FCBGA8 packaging technologies
- Thermal management support using TM1
- FSB Lane Reversal for flexible routing
- Supports C0 and C1 states only
- Execute Disable Bit support for enhanced security

This processor series represents a new family of processors designed from the groundup on a ground-breaking new low-power microarchitecture. It is manufactured on industry-leading 45-nm process with Hi-K Metal Gate technology.

This processor series enables a new class of simple and affordable internet-centric computers called "antelopes" that is best suited for applications focused on internet usage models-communicate, listen, watch, play, share, and learn.

### 1.2 Terminology

| Term | $\quad$ Definition |
| :--- | :--- |
| $\#$ | A "\#" symbol after a signal name refers to an active low signal, indicating a <br> signal is in the active state when driven to a low level. For example, when <br> RESET\# is low, a reset has been requested. Conversely, when NMI is high, <br> a nonmaskable interrupt has occurred. In the case of signals where the <br> name does not imply an active state but describes part of a binary <br> sequence (such as address or data), the "\#" symbol implies that the signal <br> is inverted. For example, D[3:0] = "HLHL" refers to a hex 'A', and D[3:0]\# <br> = "LHLH" also refers to a hex "A" (H= High logic level, L= Low logic level). |
| Front Side Bus <br> (FSB) | Refers to the interface between the processor and system core logic (also <br> known as the GMCH chipset components). |
| AGTL+ | Advanced Gunning Transceiver Logic. Used to refer to Assisted GTL+ <br> signaling technology on some Intel processors. |
| CMOS | Complementary metal-Oxide semiconductor. |
| Storage <br> Conditions | Refers to a non-operational state. The processor may be installed in a <br> platform, in a tray, or loose. Processors may be sealed in packaging or <br> exposed to free air. Under these conditions, processor landings should not <br> be connected to any supply voltages, have any I/Os biased or receive any <br> clocks. Upon exposure to "free air" (i.e., unsealed packaging or a device <br> removed from packaging material) the processor must be handled in <br> accordance with moisture sensitivity labeling (MSL) as indicated on the <br> packaging material. |
| Processor Core | Processor core die with integrated L1 and L2 cache. All AC timing and signal <br> integrity specifications are at the pads of the processor core. |
| Intel ${ }^{\circledR}$ ® 64 <br> Technology | 64 -bit memory extensions to the IA-32 architecture. <br> TDP |
| $V_{\text {CC }}$ | Thermal Design Power |
| VR | The processor core power supply |
| $V_{\text {SS }}$ | Voltage Regulator |
|  | The processor ground |

### 1.3 References

Material and concepts available in the following documents may be beneficial when reading this document:

| Document | Document Number |
| :---: | :---: |
| Intel ${ }^{\circledR}$ Atom ${ }^{\text {m }}$ Processors 200 Series Specification Update | www.intel.com/ design/processor/ specupdt/ 319978.pdf |
| Intel ${ }^{\circledR}$ Atom ${ }^{T M}$ Processors 200 Series Thermal and Mechanical Design Guidelines | www.intel.com/ design/processor/ designex/ 319979.pdf |
| AP-485, Intel® Processor Identification and CPUID Instruction Application Note | http:// <br> www.intel.com/ <br> design/processor/ <br> applnots/ <br> 241618.htm |
| Voltage Regulator-Down (VRD) 11.0 - Processor POwer Delivery Design Guidelines | http:// <br> www.intel.com/ <br> design/processor/ <br> applnots/ <br> 313214.htm |
| Intel ${ }^{\circledR} 64$ and IA-32 Architectures Software Developer's Manuals <br> Volume 1: Basic Architecture <br> Volume 2A: Instruction Set Reference, A-M <br> Volume 2B: Instruction Set Reference, $N-Z$ <br> Volume 3A: System Programming Guide <br> Volume 3B: System Programming Guide | http:// <br> www.intel.com/ products/processor/ manuals/index.htm |

## 2 Low Power Features

### 2.1 Clock Control and Low-power States

The processor supports low power states at the thread level and the package level. A thread may independently enter the C1/AutoHALT and C1/MWAIT low power states. Package low power states include Normal, Stop Grant, and Stop Grant Snoop. When both threads are in a common low-power state the central power management logic ensures the entire processor enters the respective package low power state by initiating a P_LVLx I/O read to the chipset.

The processor implements two software interfaces for requesting low power states, MWAIT instruction extensions with sub-state hints and P_LVLx reads to the ACPI P_BLK register block mapped in the processor's I/O address space. The P_LVLx I/O reads are converted to equivalent MWAIT C-state requests inside the processor and do not directly result in I/O reads on the processor FSB. The monitor address does not need to be setup before using the P_LVLx I/O read interface. The sub-state hints used for each P_LVLx read can be configured in a software programmable MSR.

Figure 2-1 shows the thread low-power states. Table 2-1 provides a mapping of thread low-power states to package low power states.

Figure 2-1. Thread Low-power States

halt break = A20M\# transition, INIT\#, INTR, NMI, PREQ\#, RESET\#, SMI\#, or APIC interrupt core state break $=$ (halt break OR Monitor event)

Table 2-1. Coordination of Thread Low-power States at the Package Level

| Thread State | Package State $^{\mathbf{2}}$ |  |
| :---: | :---: | :---: |
|  | C0 | C1 $^{\mathbf{1}}$ |
| C 0 | Normal | Normal |
| $\mathrm{C} 1^{1}$ | Normal | AutoHalt |

## NOTES:

1. AutoHALT or MWAIT/C1.
2. To enter a package state, both threads must be in a common low power state. If the threads are not in a common low power state, the package state will resolve to the highest power C state.

### 2.1.1 Thread Low-power State Descriptions

### 2.1.1.1 Thread CO State

This is the normal operating state for threads in the processor.

### 2.1.1.2 Thread C1/AutoHALT Powerdown State

C1/AutoHALT is a low-power state entered when a thread executes the HALT instruction. The processor thread will transition to the CO state upon occurrence of SMI\#, INIT\#, LINT[1:0] (NMI, INTR), or FSB interrupt messages. RESET\# will cause the processor to immediately initialize itself.

A System Management Interrupt (SMI) handler will return execution to either Normal state or the AutoHALT Powerdown state. See the InteI ${ }^{\circledR} 64$ and IA- 32 Architectures Software Developer's Manuals, Volume 3A/3B: System Programmer's Guide for more information.

While in AutoHALT Powerdown state, the processor threads will process bus snoops and snoops from the other thread. The processor will enter a snoopable sub-state (not shown in Figure 2-1) to process the snoop and then return to the AutoHALT Powerdown state.

### 2.1.1.3 Thread C1/MWAIT Powerdown State

C1/MWAIT is a low-power state entered when the processor thread executes the MWAIT(C1) instruction. Processor behavior in the MWAIT state is identical to the AutoHALT state except that Monitor events can cause the processor to return to the C0 state. See the Intel ${ }^{\circledR} 64$ and IA-32 Architectures Software Developer's Manuals, Volume 2A: Instruction Set Reference, A-M and Volume 2B: Instruction Set Reference, $\mathrm{N}-\mathrm{Z}$, for more information.

### 2.1.2 Package Low-power State Descriptions

Note: $\quad$ The following state descriptions assume that both threads are in the a common low power state. For cases when only one thread is in a low power state, please see Section 2.1.1.

### 2.1.2.1 Normal State

This is the normal operating state for the processor. The processor remains in the Normal state when the threads are in the C0, C1/AutoHALT, or C1/MWAIT state.

### 2.1.3 Front Side Bus

The processor has only one signaling mode, where the data and address busses and the strobe signals are operating in GTL mode. The reason to use GTL is to improve signal integrity.

## 3 Electrical Specifications

This chapter contains signal group descriptions, absolute maximum ratings, voltage identification and power sequencing. The chapter also includes DC and AC specifications, including timing diagrams.

### 3.1 FSB and GTLREF

The processor supports two kinds of signalling protocol: Complementary Metal Oxide Semiconductor (CMOS), and Advanced Gunning Transceiver Logic (AGTL+). For FSB data and address bus, only AGTL+ is used.

The termination voltage level for the processor CMOS and AGTL+ signals is $\mathrm{V}_{\mathrm{T}}=1.10$ V (nominal). Due to speed improvements to data and address bus, signal integrity and platform design methods have become more critical than with previous processor families.

The CMOS sideband signals are listed in Table 3-5.
The AGTL+ inputs, including the sideband signals listed in Table 3-5, require a reference voltage (GTLREF_MA, GTLREF_EA) that is used by the receivers to determine if a signal is a logical 0 or a logical 1. GTLREF_* must be generated on the system board. Termination resistors are provided on the processor silicon and are terminated to its I/O voltage ( $\mathrm{V}_{\mathrm{T}}$ ). The appropriate chipset will also provide on-die termination, thus eliminating the need to terminate the bus on the system board for most AGTL+ signals.

The AGTL+ bus depends on incident wave switching. Timing calculations for AGTL+ signals are based on flight time as opposed to capacitive deratings. Analog signal simulation of the FSB, including trace lengths, is highly recommended when designing a system.

### 3.2 Power and Ground Pins

For clean, on-chip power distribution, the processor will have a large number of $\mathrm{V}_{\Pi}$ (FSB AGTL+ reference voltage), $\mathrm{V}_{\mathrm{CCP}}$ (power) and $\mathrm{V}_{\mathrm{SS}}$ (ground) inputs. All power pins must be connected to $\mathrm{V}_{\mathrm{CCP}}$ power planes while all $\mathrm{V}_{\mathrm{SS}}$ pins must be connected to system ground planes. Use of multiple power and ground planes is recommended to reduce I*R drop. The processor $\mathrm{V}_{\mathrm{CCP}}$ pins must be supplied the voltage stated in Table 3-7

### 3.3 Decoupling Guidelines

Due to its large number of transistors and high internal clock speeds, the processor is capable of generating large average current swings between low and full power states. This may cause voltages on power planes to sag below their minimum values if bulk decoupling is not adequate. Larger bulk storage, such as electrolytic capacitors, supply current during longer lasting changes in current demand by the component, such as coming out of an idle condition. Similarly, they act as a storage well for current when entering an idle condition from a running condition. Care must be taken in the board design to ensure that the voltage provided to the processor remains within the specifications listed in Table 3-7. Failure to do so can result in timing violations or reduced lifetime of the component. For further information and design guidelines, refer to the Voltage Regulator-Down (VRD) 11.0 Processor Power Delivery Design Guidelines.

### 3.3.1 $\quad V_{\text {CCP }}$ Decoupling

$\mathrm{V}_{\mathrm{CC}}$ regulator solutions need to provide bulk capacitance with a low Effective Series Resistance (ESR) and keep a low interconnect resistance from the regulator to the socket. Bulk decoupling for the large current swings when the part is powering on, or entering/exiting low-power states, must be provided by the voltage regulator solution For more details on decoupling recommendations, refer to the Voltage Regulator-Down (VRD) 11.0 Processor Power Delivery Design Guidelines.

### 3.3.2 FSB AGTL+ Decoupling

The processors integrate signal termination on the die. Decoupling must also be provided by the system motherboard for proper AGTL+ bus operation.

Electrical Specifications

### 3.4 Voltage Identification and Power Sequencing

Table 3-2. Voltage Identification Definition

| VID | VID | VID | VID | VID | VID |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ |  |  |
| $\mathbf{1}$ | $\mathbf{0}$ | VCC <br> $\mathbf{( V )}$ |  |  |  |  |  |
| 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1.2000 |
| 0 | 1 | 0 | 0 | 0 | 1 | 0 | 1.1875 |
| 0 | 1 | 0 | 0 | 0 | 1 | 1 | 1.1750 |
| 0 | 1 | 0 | 0 | 1 | 0 | 0 | 1.1625 |
| 0 | 1 | 0 | 0 | 1 | 0 | 1 | 1.1500 |
| 0 | 1 | 0 | 0 | 1 | 1 | 0 | 1.1375 |
| 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1.1250 |
| 0 | 1 | 0 | 1 | 0 | 0 | 0 | 1.1125 |
| 0 | 1 | 0 | 1 | 0 | 0 | 1 | 1.1000 |
| 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1.0875 |
| 0 | 1 | 0 | 1 | 0 | 1 | 1 | 1.0750 |
| 0 | 1 | 0 | 1 | 1 | 0 | 0 | 1.0625 |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1.0500 |
| 0 | 1 | 0 | 1 | 1 | 1 | 0 | 1.0375 |
| 0 | 1 | 0 | 1 | 1 | 1 | 1 | 1.0250 |
| 0 | 1 | 1 | 0 | 0 | 0 | 0 | 1.0125 |
| 0 | 1 | 1 | 0 | 0 | 0 | 1 | 1.0000 |
| 0 | 1 | 1 | 0 | 0 | 1 | 0 | 0.9875 |
| 0 | 1 | 1 | 0 | 0 | 1 | 1 | 0.9750 |
| 0 | 1 | 1 | 0 | 1 | 0 | 0 | 0.9625 |
| 0 | 1 | 1 | 0 | 1 | 0 | 1 | 0.9500 |


| VID | VID | VID | VID | VID | VID | VID | VCC (V) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |  |
| 0 | 1 | 1 | 0 | 1 | 1 | 0 | 0.9375 |
| 0 | 1 | 1 | 0 | 1 | 1 | 1 | 0.9250 |
| 0 | 1 | 1 | 1 | 0 | 0 | 0 | 0.9125 |
| 0 | 1 | 1 | 1 | 0 | 0 | 1 | 0.9000 |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | 0.8875 |
| 0 | 1 | 1 | 1 | 0 | 1 | 1 | 0.8750 |
| 0 | 1 | 1 | 1 | 1 | 0 | 0 | 0.8625 |
| 0 | 1 | 1 | 1 | 1 | 0 | 1 | 0.8500 |
| 0 | 1 | 1 | 1 | 1 | 1 | 0 | 0.8375 |
| 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0.8250 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0.8125 |
| 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0.8000 |
| 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0.7875 |
| 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0.7750 |
| 1 | 0 | 0 | 0 | 1 | 0 | 0 | 0.7625 |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 0.7500 |
| 1 | 0 | 0 | 0 | 1 | 1 | 0 | 0.7375 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 0.7250 |
| 1 | 0 | 0 | 1 | 0 | 0 | 0 | 0.7125 |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 0.7000 |

The VID specification for the processor is defined by the RS - Voltage Regulator-Down (VRD) 11.0 Processor Power Delivery Design Guidelines.

The processor uses seven voltage identification pins,VID[6:0], to support automatic selection of power supply voltages. The VID pins for processor are CMOS outputs driven by the processor VID circuitry. Table 3-2 specifies the voltage level corresponding to the state of VID[6:0]. A " 1 " in this refers to a high-voltage level and a " 0 " refers to low-voltage level. For more details about VR design to support the processor power supply requirements, Refer to the Voltage Regulator-Down (VRD) 11.0 Processor Power Delivery Design Guidelines.

VRD11 has 8 VID pins (VID[7:0]) compared to 7 VID pins for processor. VRD11 VID[n] should be connected to processor VID[n-1]. VRD11 VID[0] should be tied to Vss.

Table 3-3. Processor VID Pin to VRD11 VID Pin Mapping

| Processor VID pin | map to VRD11 VID pin |
| :---: | :---: |
| 6 | 7 |
| 5 | 6 |
| 4 | 5 |
| 3 | 4 |
| 2 | 3 |
| 1 | 2 |
| 0 | 1 |
|  | 0 (tie to $V_{\mathrm{SS}}$ ) |

Power source characteristics must be stable whenever the supply to the voltage regulator is stable.

### 3.5 Catastrophic Thermal Protection

The processor supports the THERMTRIP\# signal for catastrophic thermal protection. An external thermal sensor should also be used to protect the processor and the system against excessive temperatures. Even with the activation of THERMTRIP\#, which halts all processor internal clocks and activity, leakage current can be high enough such that the processor cannot be protected in all conditions without the removal of power to the processor. If the external thermal sensor detects a catastrophic processor temperature of $125^{\circ} \mathrm{C}$ (maximum), or if the THERMTRIP\# signal is asserted, the $\mathrm{V}_{\mathrm{CCP}}$ supply to the processor must be turned off within 500 ms to prevent permanent silicon damage due to thermal runaway of the processor. THERMTRIP\# functionality is not ensured if the PWRGOOD signal is not asserted.

### 3.6 Reserved and Unused Pins

All RESERVED (RSVD) pins must remain unconnected. Connection of these pins to $\mathrm{V}_{\mathrm{CC}}$, $\mathrm{V}_{\mathrm{SS}}$, or to any other signal (including each other) can result in component malfunction or incompatibility with future processors. See section Chapter 4.2 for a pin listing of the processor and the location of all RSVD pins.

For reliable operation, always connect unused inputs or bidirectional signals to an appropriate signal level. Unused active low AGTL+ inputs may be left as no connects if AGTL+ termination is provided on the processor silicon. Unused active high inputs should be connected through a resistor to ground ( $\mathrm{V}_{\mathrm{SS}}$ ). Unused outputs can be left unconnected.

### 3.7 FSB Frequency Select Signals (BSEL[2:0])

Only 133 MHz is supported by the processor. The BSEL[2:0] signals need to be set accordingly to select the frequency of the processor input clock (BCLK[1:0]). These signals should be connected to the clock chip and the appropriate chipset on the platform. The BSEL encoding for BCLK[1:0] is shown in Table 3-4.

Table 3-4. BSEL[2:0] Encoding for BCLK Frequency

| BSEL[2] | BSEL[1] | BSEL[0] | BCLK Frequency |
| :---: | :---: | :---: | :---: |
| L | L | H | 133 MHz |

NOTE: All other bus selections reserved.

### 3.8 FSB Signal Groups

To simplify the following discussion, the FSB signals have been combined into groups by buffer type. AGTL+ input signals have differential input buffers, which use GTLREF as a reference level. In this document, the term "AGTL+ Input" refers to the AGTL+ input group as well as the AGTL+ I/O group when receiving. Similarly, "AGTL+ Output" refers to the AGTL+ output group as well as the AGTL+ I/O group when driving.

With the implementation of a source synchronous data bus comes the need to specify two sets of timing parameters. One set is for common clock signals which are dependent upon the rising edge of BCLKO (ADS\#, HIT\#, HITM\#, etc.) and the second set is for the source synchronous signals which are relative to their respective strobe lines (data and address) as well as the rising edge of BCLKO. Asychronous signals are still present (A20M\#, IGNNE\#, etc.) and can become active at any time during the clock cycle. Table 3-5 identifies which signals are common clock, source synchronous, and asynchronous.

Table 3-5. FSB Pin Groups

| Signal Group | Type | Signals1 |
| :---: | :---: | :---: |
| AGTL+ Common Clock Input | Synchronous to BCLK[1:0] | BPRI\#, DEFER\#, PREQ\#, RESET\#, RS[2:0]\#, TRDY\#, DPWR\# |
| AGTL+ Common Clock I/O | Synchronous to BCLK[1:0] | $\begin{aligned} & \text { ADS\#, BNR\#, BPM[3:0]\#, BRO\#, DBSY\#, DRDY\#, } \\ & \text { HIT\#, HITM\#, LOCK\#, PRDY\#, BPM_2[3:0], } \end{aligned}$ |
| AGTL+ Source Synchronous I/O | Synchronous to assoc. strobe | Signals Associated Strobe <br> REQ[4:0]\#, A[16:3]\# ADSTB0\# <br> A[32:17]\# ADSTB1\# <br> D[15:0]\#, DBIO\#, DINV[0]\# DSTBPO\#, DSTBNO\# <br> D[31:16]\#, DBI1\#, DINV[0]\# DSTBP1\#, DSTBN1\# <br> D[47:32]\#, DBI2\#, DINV[0]\# DSTBP2\#, DSTBN2\# <br> D[63:48]\#, DBI3\#, DINV[0]\# DSTBP3\#, DSTBN3\# |
| AGTL+ Strobes | Synchronous to BCLK[1:0] | ADSTB[1:0]\#, DSTBP[3:0]\#, DSTBN[3:0]\# |
| CMOS Input | Asynchronous | IGNNE\#, INIT\#, LINTO/INTR, LINT1/NMI, PWRGOOD, SMI\# |
| Open Drain Output | Asynchronous | FERR\#, THERMTRIP\#, IERR\# |
| Open Drain I/O | Asynchronous | PROCHOT\# |
| CMOS Output | Asynchronous | VID[6:0], BSEL[2:0] |
| CMOS Input | Synchronous to TCK | TCK, TDI, TMS, TRST\#, TDI_M |
| Open Drain Output | Synchronous to TCK | TDO, TDO_M |
| FSB Clock | Clock | BCLK[1:0] |
| Power/Other |  | COMP[3:0], GTLREF_MA, GTLREF_EA, TEST2/DCLK, TEST1/ACLK, THERMDA, THERMDC, $\mathrm{V}_{\mathrm{CCA}}, \mathrm{V}_{\mathrm{CCP}}, \mathrm{V}_{\mathrm{TT}}$, $\mathrm{V}_{\text {CC_SENSE }}, \mathrm{V}_{\text {SS }}, \mathrm{V}_{\text {SS_SENSE }}, \mathrm{V}_{\text {CCQ }}[1: 0]$, COMP_2[3:0], THRMDA_2, THRMDC_2 |

## NOTES:

1. Refer to Chapter 4 for signal descriptions and termination requirements.
2. In processor systems where there is no debug port implemented on the system board, these signals are used to support a debug port interposer. In systems with the debug port implemented on the system board, these signals are no connects.
3. PROCHOT\# signal type is open drain output and CMOS input.
4. On die termination differs from other AGTL+ signals, refer to your Platform Design Guidelines for up to day recommendations.

### 3.9 CMOS Asynchronous Signals

CMOS input signals are shown in Table 3-5. Legacy output FERR\#, IERR\# and other non- AGTL+ signals (THERMTRIP\# and PROCHOT\#) use Open Drain output buffers. These signals do not have setup or hold time specifications in relation to BCLK[1:0]. However, all of the CMOS signals are required to be asserted for more than 4 BCLKs for the processor to recognize them. See Section 3.11 for the DC specifications for the CMOS signal groups.

### 3.10 Maximum Ratings

Table 3-6 specifies absolute maximum and minimum ratings. Within functional operation limits, functionality and long-term reliability can be expected.

At conditions outside functional operation condition limits, but within absolute maximum and minimum ratings, neither functionality nor long term reliability can be expected. If a device is returned to conditions within functional operation limits after having been subjected to conditions outside these limits, but within the absolute maximum and minimum ratings, the device may be functional, but with its lifetime degraded depending on exposure to conditions exceeding the functional operation condition limits.

At conditions exceeding absolute maximum and minimum ratings, neither functionality nor long term reliability can be expected. Moreover, if a device is subjected to these conditions for any length of time then, when returned to conditions within the functional operating condition limits, it will either not function or its reliability will be severely degraded.

Although the processor contains protective circuitry to resist damage from static electric discharge, precautions should always be taken to avoid high static voltages or electric fields.

Table 3-6. Processor Absolute Maximum Ratings

| Symbol | Parameter | Min | Max | Unit | Notes ${ }^{\mathbf{1 , 5}}$ |
| :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{T}_{\text {STORAGE }}$ | Processor Storage Temperature | -40 | 85 | ${ }^{\circ} \mathrm{C}$ | $2,3,4$ |
| $\mathrm{~V}_{\text {CCP }}$ | Any Processor Supply Voltage with Respect to $\mathrm{V}_{\text {SS }}$ | -0.3 | 1.1 | V | 6 |
| $\mathrm{~V}_{\text {CCA }}$ | PLL power supply | -0.3 | 1.575 | V |  |
| VinAGTL+ | AGTL+ Buffer DC Input Voltage with Respect to $\mathrm{V}_{\mathrm{SS}}$ | -0.1 | 1.1 | V |  |
| VinAsynch_CMOS | CMOS Buffer DC Input Voltage with Respect to $\mathrm{V}_{\mathrm{SS}}$ | -0.1 | 1.1 | V |  |

## NOTES:

1. For functional operation, all processor electrical, signal quality, mechanical and thermal specifications must be satisfied.
2. Storage temperature is applicable to storage conditions only. In this scenario, the processor must not receive a clock, and no lands can be connected to a voltage bias. Storage within these limits will not affect the long term reliability of the device. For functional operation, refer to the processor case temperature specifications.
3. This rating applies to the processor and does not include any tray or packaging.
4. Failure to adhere to this specification can affect the long term reliability of the processor.
5. The $\mathrm{V}_{\mathrm{CC}}$ max supported by the process is 1.2 V but the parameter can change (burnin voltage is higher).

### 3.11 Processor DC Specifications

The processor DC specifications in this section are defined at the processor core (pads) unless noted otherwise. See Chapter 4 for the pin signal definitions and signal pin assignments. Most of the signals on the FSB are in the AGTL+ signal group. The DC specifications for these signals are listed in Table 3-9. DC specifications for the CMOS group are listed in Table 3-10.

Table 3-9 through Table 3-11 list the DC specifications for the processor and are valid only while meeting specifications for junction temperature, clock frequency, and input voltages. Unless specified otherwise, all specifications for the processor are at $\mathrm{T}_{\mathrm{J}}=$ $90^{\circ} \mathrm{C}$. Care should be taken to read all notes associated with each parameter.

Table 3-7. Voltage and Current Specifications for the processor

| Symbol |  | Parameter | Min | Тур | Max | Unit | Notes ${ }^{4}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| FSB <br> Frequency | BCLK Frequency |  | 132.6 | 133.33 | 133.5 | MHz |  |
| $\mathrm{V}_{\text {T }}$ | FSB AGTL+ termination voltage with respect to $V_{S S}$ |  | 1.05 | 1.10 | 1.15 | V |  |
| $\mathrm{V}_{\text {CCP }}$ | Vcc Core voltage with respect to $\mathrm{V}_{\text {SS }}$ |  | 0.7 |  | 1.20 | V | 6 |
| $\mathrm{V}_{\text {CC }}$ воот | Default $\mathrm{V}_{\mathrm{CC}}$ voltage for initial power up |  | 1.10 |  | 1.20 | V | 8 |
| $\mathrm{V}_{\text {CCA }}$ | PLL Supply voltage |  | 1.425 | 1.5 | 1.575 | V |  |
| $\mathrm{I}_{\mathrm{T}}$ | $\mathrm{I}_{\mathrm{CC}}$ for $\mathrm{V}_{\text {TT }}$ supply after $\mathrm{V}_{\mathrm{CC}}$ stable $\mathrm{I}_{\mathrm{CC}}$ for $\mathrm{V}_{\mathrm{T}}$ supply at startup |  |  |  | $\begin{aligned} & \hline 1.5 \\ & 2.5 \end{aligned}$ | A | 3 |
| $\mathrm{I}_{\text {CCDES }}$ | $\mathrm{I}_{\mathrm{CC}}$ for processors recommended design target (estimated) |  |  |  | 8 | A | 3 |
| $\mathrm{I}_{\mathrm{CC}}$ | $\mathrm{I}_{\text {CC }}$ for processors |  |  |  |  |  |  |
|  | Processor Number | Core Frequency/Voltage |  |  |  |  |  |
|  | 330 | 1.6 GHz @ $\mathrm{V}_{\mathrm{CC}}$ (AVID controlled) |  |  | 8 | A | 1, 2 |
| $\mathrm{I}_{\text {AH, }}$ | ICC Auto-Halt |  |  |  | 4 | A | 1,2 |
| I CCA | $\mathrm{I}_{\text {CC }}$ for $\mathrm{V}_{\text {CCA }}$ Supply |  |  |  | 260 | mA |  |

## NOTES:

1. Specified at $90^{\circ} \mathrm{C} \mathrm{T}_{J}$.
2. Specified at the nominal $\mathrm{V}_{\mathrm{CC}}$.
3. Refer to the RS - Voltage Regulator-Down (VRD) 11.0-Processor Power Delivery Design Guidelines for design target capability.
4. Unless otherwise noted, all specifications in this table are based on estimates and simulations or empirical data. These specifications will be updated with characterized data from silicon measurements at a later date.
5. This is the Vcc range, not the absolute Vcc for the core. The Vccp tolerance should be $+/-50 \mathrm{mV}$, inclusive of ripple, VR tolerance and transient (droop and overshoot).
6. Since processor is soldered down with no loadline and no dynamic VID, there is no "socket load line slope (SKT_LL)"; no "socket load line tolerance band" but only "Tolerance Band (TOB)" of 50 mV ; no "maximum overshoot above VID (OS_AMP)"; no "maximum overshoot time duration above VID (OS_TIME)"; no "peak to peak ripple amplitude (RIPPLE)"; no "thermal compensation voltage drift (THERMAL_DRIFT)"; no "Maximum DC test (Current I_DC_MAX)"; no "minimum DC test *(Current I_DC_MIN)"; "Voltage Regulator Thermal Design Current (VR_TDC) of 3.64A; "current step rise time (I_RISE) of 5A/us"
7. $\pm 50 \mathrm{mV}$ tolerance

Table 3-8. FSB Differential BCLK Specifications

| Symbol | Parameter | Min | Typ | Max | Unit | Figure | Notes $^{\mathbf{1}}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage |  |  | 1.15 | V |  | 7,8 |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input Low Voltage |  |  | -0.3 | V |  | 7,8 |
| $\mathrm{~V}_{\mathrm{CROss}}$ | Crossing Voltage | 0.3 |  | 0.55 | V |  | $2,7,9$ |
| $\Delta \mathrm{~V}_{\mathrm{CROss}}$ | Range of Crossing Points |  |  | 140 | mV |  | $2,7,5$ |
| $\mathrm{~V}_{\mathrm{SWING}}$ | Differential Output Swing | 300 |  |  | mV |  | 6 |
| $\mathrm{I}_{\mathrm{LI}}$ | Input Leakage Current | -5 |  | +5 | $\mu \mathrm{~A}$ |  | 3 |
| Cpad | Pad Capacitance | 1.2 | 1.45 | 2.0 | pF |  | 4 |

## NOTES:

1. Unless otherwise noted, all specifications in this table apply to all processor frequencies.
2. Crossing Voltage is defined as absolute voltage where rising edge of BCLKO is equal to the falling edge of BCLK1.
3. For Vin between 0 V and $\mathrm{V}_{\mathrm{IH}}$.
4. Cpad includes die capacitance only. No package parasitics are included.
5. $\quad \Delta \mathrm{V}_{\text {CROss }}$ is defined as the total variation of all crossing voltages as defined in note 2.
6. Measurement taken from differential waveform.
7. Measurement taken from single-ended waveform.
8. "Steady state" voltage, not including Overshoots or Undershoots.
9. Only applies to the differential rising edge (clock rising and clock\# falling).

Table 3-9. AGTL+/CMOS Signal Group DC Specifications

| Symbol | Parameter | Min | Тур | Max | Unit | Notes ${ }^{1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {T }}$ | I/O Voltage | 1.05 | 1.10 | 1.15 | V | 11 |
| GTLREF_EA | GTL Reference Voltage, endagent |  | $0.63 \mathrm{~V}_{T}$ |  | V | $\begin{gathered} 5,12 \\ 13 \end{gathered}$ |
| GTLREF_MA | GTL Reference Voltage, middle-agent |  | $0.58 \mathrm{~V}_{T}$ |  | V | $\begin{gathered} 5,12, \\ 13 \end{gathered}$ |
| $\mathrm{R}_{\text {COMP }}$ | Compensation Resistor COMP[0] \& COMP[2] <br> COMP[1] \& COMP[3] | $\begin{gathered} 24.75 \\ 49.5 \end{gathered}$ | $\begin{aligned} & 25 \\ & 50 \end{aligned}$ | $\begin{gathered} 25.25 \\ 50.5 \end{gathered}$ | $\Omega$ | 9, 11 |
| $\mathrm{R}_{\text {ODT }}$ | Termination Resistor |  | 55 |  | $\Omega$ | 10 |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage | GTLREF+0.10 | $\mathrm{V}_{\text {T }}$ | $\mathrm{V}_{\text {TT }}+0.10$ | V | 3,5 |
| $\mathrm{V}_{\text {IL }}$ | Input Low Voltage | -0.10 | 0 | GTLREF-0.10 | V | 2,4 |
| $\mathrm{V}_{\mathrm{OH}}$ | Output High Voltage | $\mathrm{V}_{T \mathrm{~T}^{-0.10}}$ | $\mathrm{V}_{T}$ | $V_{T}$ | V | 5 |
| $\mathrm{R}_{\text {TT }}$ | Termination Resistance | 45 | 50 | 55 | $\Omega$ | 6 |
| $\begin{gathered} \mathrm{R}_{\mathrm{ON}}(\mathrm{GTL} \\ \text { mode) } \end{gathered}$ | GTL Buffer on Resistance |  | 8.5 |  | $\Omega$ |  |
| $\mathrm{I}_{\text {LI }}$ | Input Leakage Current |  |  | $\pm 100$ | $\mu \mathrm{A}$ | 7 |
| Cpad | Pad Capacitance | 1.8 | 2.1 | 2.75 | pF | 8 |

## NOTES:

1. Unless otherwise noted, all specifications in this table apply to all processor frequencies.
2. $\quad \mathrm{V}_{\mathrm{IL}}$ is defined as the maximum voltage level at a receiving agent that will be interpreted as a logical low value.
3. $\quad \mathrm{V}_{\mathrm{IH}}$ is defined as the minimum voltage level at a receiving agent that will be interpreted as a logical high value.
4. $\quad \mathrm{V}_{\mathrm{IH}}$ and $\mathrm{V}_{\mathrm{OH}}$ may experience excursions above $\mathrm{V}_{\mathrm{T}}$. However, input signal drivers must comply with the signal quality specifications.
5. GTLREF should be generated from $V_{T}$ with a $1 \%$ tolerance resistor divider. The $V_{T}$ referred to in these specifications is the instantaneous $\mathrm{V}_{\mathrm{T}}$.
6. $\quad R_{T}$ is the on-die termination resistance measured at $V_{O L}$ of the AGTL+ output driver. Measured at $0.31 * V_{C C P}$. $R_{T T}$ is connected to $V_{C C P}$ on die. Refer to processor I/O buffer models for I/V characteristics.
7. $\quad$ Specified with on die $R_{T}$ and $R_{O N}$ are turned off. Vin between 0 and $V_{T T}$.
8. Cpad includes die capacitance only. No package parasitics are included.
9. This is the external resistor on the comp pins.
10. On die termination resistance, measured at $0.33 * \mathrm{~V}_{\mathrm{T}}$.
11. $R_{\text {COMP }}$ resistance must be provided on the system board with $1 \%$ resistors.
12. GTLREF_EA \& GTLREF_MA resistor divider needs to be separate. This is for better FSB margin.

Table 3-10.Legacy CMOS Signal Group DC Specifications

| Symbol | Parameter | Min | Typ | Max | Unit | Notes $^{\mathbf{1}}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\Pi}$ | I/O Voltage | 1.05 | 1.10 | 1.15 | V | 9 |
| $\mathrm{~V}_{\mathrm{IH}}$ | Input High Voltage | $0.7 * \mathrm{~V}_{\Pi}$ | $\mathrm{V}_{\Pi}$ | $\mathrm{V}_{\Pi}+0.1$ | V | 2 |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input Low Voltage CMOS | -0.10 | 0.00 | $0.3^{*} \mathrm{~V}_{\Pi}$ | V | 2,3 |
| $\mathrm{~V}_{\mathrm{OH}}$ | Output High Voltage | $0.9 * \mathrm{~V}_{\Pi}$ | $\mathrm{V}_{\Pi}$ | $\mathrm{V}_{\Pi}+0.1$ | V | 2 |
| $\mathrm{~V}_{\mathrm{OL}}$ | Output Low Voltage | -0.10 | 0 | $0.1^{*} \mathrm{~V}_{\Pi}$ | V | 2 |
| $\mathrm{I}_{\mathrm{OH}}$ | Output High Current | 1.5 |  | 4.1 | mA | 5 |
| $\mathrm{I}_{\mathrm{OL}}$ | Output Low Current | 1.5 |  | 4.1 | mA | 4 |
| $\mathrm{I}_{\mathrm{LI}}$ | Input Leakage Current |  |  | $\pm 100$ | $\mu \mathrm{~A}$ | 6 |
| Cpad1 | Pad Capacitance | 1.6 | 2.1 | 2.55 | pF | 7 |
| Cpad2 | Pad Capacitance for CMOS <br> Input | 0.95 | 1.2 | 1.45 |  | 8 |

## NOTES:

1. Unless otherwise noted, all specifications in this table apply to all processor frequencies.
2. The $\mathrm{V}_{\mathrm{T}}$ referred to in these specifications refers to instantaneous $\mathrm{V}_{\mathrm{T}}$.
3. Measured at $0.1^{*} \mathrm{~V}_{\Pi}$.
4. Measured at $0.9 * V_{T \Gamma}$.
5. For Vin between 0 V and $\mathrm{V}_{\mathrm{T}}$. Measured when the driver is tri-stated.
6. Cpad1 includes die capacitance only for DPRSTP\#, DPSLP\#, PWRGOOD. No package parasitics are included.
7. Cpad2 includes die capacitance for all other CMOS input signals. No package parasitics are included.

Table 3-11.Open Drain Signal Group DC Specifications

| Symbol | Parameter | Min | Typ | Max | Unit | Notes $^{\mathbf{1}}$ |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{OH}}$ | Output High Voltage | $\mathrm{V}_{\Pi \text { T }}{ }^{5} \%$ | $\mathrm{~V}_{\Pi}$ | $\mathrm{V}_{\Pi+}+5 \%$ | V | 3 |
| $\mathrm{~V}_{\mathrm{OL}}$ | Output Low Voltage | 0 |  | 0.20 | V |  |
| $\mathrm{I}_{\mathrm{OL}}$ | Output Low Current | 16 |  | 50 | mA | 2 |
| $\mathrm{I}_{\mathrm{LO}}$ | Output Leakage Current |  |  | $\pm 200$ | $\mu \mathrm{~A}$ | 4 |
| Cpad | Pad Capacitance | 1.9 | 2.2 | 2.45 | pF | 5 |

## NOTES:

1. Unless otherwise noted, all specifications in this table apply to all processor frequencies.
2. Measured at 0.2 V .
3. $\quad \mathrm{V}_{\mathrm{OH}}$ is determined by value of the external pull-up resistor to $\mathrm{V}_{\mathrm{T}}$. Refer to platform design guide for details.
4. For Vin between 0 V and $\mathrm{V}_{\mathrm{OH}}$.
5. Cpad includes die capacitance only. No package parasitics are included.
3.12 AGTL+ FSB Specifications

Routing topology recommendations may be found in the appropriate platform design guides. Termination resistors are not required for most AGTL+ signals, as these are integrated into the processor silicon.

Valid high and low levels are determined by the input buffers which compare a signal's voltage with a reference voltage called GTLREF (known as $\mathrm{V}_{\text {REF }}$ in previous documentation).

Table 3-9 lists the GTLREF specifications. The AGTL+ reference voltage (GTLREF) should be generated on the system board using high precision voltage divider circuits. It is important that the system board impedance is held to the specified tolerance, and that the intrinsic trace capacitance for the AGTL+ signal group traces is known and well- controlled. For more details on platform design, see the appropriate platform design guides.

Package Mechanical Specifications and Ball Information

## 4 Package Mechanical Specifications and Ball I nformation

This chapter provides the package specifications, pinout assignments, and signal description.

### 4.1 Package Mechanical Specifications

The processor will be available in $512 \mathrm{kB}, 437$ pins in FCBGA8 package.

### 4.1.1 Package Mechanical Drawings

Figure 4-2. Package Mechanical Drawing


### 4.1.2 Package Loading Specifications

Package loading is 10 lb max static compressive.

### 4.1.3 Processor Mass Specifications

Processor mass is 1.8 g .

### 4.2 Processor Pinout Assignment

Figure 4-3 and Figure 4-4 are graphic representations of the processor pinout assignments. Table 4-12 lists the pinout by signal name.

Package Mechanical Specifications and Ball Information

Figure 4-3. Pinout Diagram (Top View, Left Side)

|  | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 | 13 | 14 | A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A |  | VSS | RSVD | VSS | D[54]\# | D[56]\# | $\underset{\text { GTLREF_ }}{\substack{\text { MA }}}$ | VSS | VCCQ0 | VCCP | VCCP | VCCP | RSVD | $\begin{gathered} \mathrm{A}[35] \\ \# \end{gathered}$ |  |
| B | VSS | VSS | D[60]\# | D[52]\# | VSS | D[59]\# | CMREF | vSS | vCCQ0 | VCCP | VCCP | VCCP | VSS | $\begin{gathered} \mathrm{A}[33] \\ \# \end{gathered}$ | B |
| c | RSVD | D[48]\# | D[55]\# | D[61]\# | DINV[3] | D[58]\# | D[62]\# | vSS | VTT | VCCP | VCCP | VCCP | $\begin{aligned} & \text { VCCSE } \\ & \text { NSE } \end{aligned}$ | $\mathrm{A}[22]$ | c |
| D | vSS | D[63]\# | D[51]\# | RSVD | VSS | GTLREF_EA | VCCA | vss | VTT | VCCP | VCCP | VCCP | $\begin{aligned} & \text { VSSSE } \\ & \text { NSE } \end{aligned}$ | vSS | D |
| E | D[53]\# | DSTBN[3] | VSS | THRMDA | THRMDC | VSS | VSS | VSS | VTT | VCCP | VCCP | VCCP | VTT | VTT | E |
| F | D[50]\# | D[57]\# | DSTBP[3] | THRMDC_2 | VSS | VSS | VSS | VTT | VTT | VCCP | VCCP | VCCP | VTT | VTT | F |
| G | VSS | D[49]\# | D[40]\# | THRMDA_2 | BSEL[2] | BPM_2[0] | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | $\begin{array}{\|c\|} \hline \text { VCCQ } \\ 1 \end{array}$ | G |
| H | D[46]\# | D[41]\# | VSS | VSS | BSEL[1] | BPM2[1] | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | $\begin{array}{\|c} \text { VCCQ } \\ 1 \end{array}$ | H |
| J | D[47]\# | D[45]\# | D[38]\# | IGNNE\# | VSS | BSEL[0] | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | J |
| K | VSS | DSTBN[2] | DSTBP[2] | BPM_2[2] | BPM_2[3] | VSS | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | k |
| L | DINV[2] | D[43]\# | VSS | VSS | COMP_2[0] | VSS | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | L |
| M | VSS | D[36]\# | D[44]\# | RSVD | VSS | EXTBGREF | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | M |
| N | D[35]\# | D[42]\# | D[39]\# | COMP_2[2] | COMP_2[1] | RSVD | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | N |
| P | D[34]\# | D[37] \# | VSS | COMP_2[3] | RSVD | VSS | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | P |
| R | VSS | D[33]\# | D[32]\# | RSVD | VSS | RSVD | VSS | VTT | VSS | VCCP | VCCP | VCCP | VSS | VTT | R |
| T | COMP[0] | COMP[1] | D[28]\# | VSS | RSVD | RSVD | VSS | VTT | VSS | VSS | VSS | VSS | VSS | VTT | T |
| U | D[19]\# | D[27]\# | VSS | DPWR\# | RSVD | VSS | VSS | VTT | VTT | VTT | VTT | VTT | VTT | VTT | U |
| v | VSS | D[30]\# | D[26]\# | VSS | RSVD | VSS | VSS | VSS | RSVD | VTT | BCLK[0] | BCLK[1] | VSS | VSS | v |
| w | VSS | D[25]\# | D[18]\# | D[31]\# | VSS | D[21]\# | D[20]\# | VSS | D[15]\# | D[1]\# | VSS | D[5]\# | D[13]\# | VSS | w |
| Y | VSS | VSS | D24[]\# | DSTBN[1] | DSTBP[1] | DINV[1] | D[22]\# | D[17]\# | D[8]\# | D[7]\# | D[0]\# | D[2]\# | D[9]\# | $\begin{array}{\|l} \text { DSTB } \\ \text { N[0] } \end{array}$ | Y |
| AA |  | VSS | VSS | VSS | D[16]\# | D[23]\# | VSS | D[29]\# | D[14]\# | VSS | D[4]\# | VSS | D[11]\# | D[3]\# | AA |
|  | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 | 13 | 14 |  |

Figure 4-4. Pinout Diagram (Top View, Right Side)

A
B
C
D
E

F
G
H
J
к

| 15 | 16 | 17 | 18 | 19 | 20 | 21 | A |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VSS | A[20]\# | A[32]\# | VSS | RSVD | VSS |  |  |
| A[34]\# | A[29]\# | A[30]\# | A[27] \# | ADSTB[1]\# | VSS | VSS | B |
| A[28]\# | A[31]\# | VSS | A[23]\# | A[17]\# | A[24]\# | RSVD | C |
| RESET\# | VID[1] | RSVD | VSS | A[21]\# | A[26]\# | VSS | D |
| VSS | VSS | VID[5] | VID[2] | VSS | A[25]\# | A[19]\# | E |
| VID[0] | IERR\# | VSS | VSS | A[18]\# | COMP[2] | COMP[3] | F |
| VID[3] | VID[4] | PROCHOT\# | VID[6] | REQ[2]\# | A[9]\# | VSS | G |
| BPM[2]\# | VSS | THERMTRIP\# | VSS | VSS | A[4]\# | A[11]\# | H |
| BPM[3] \# | PREQ\# | VSS | BPM[1] \# | A[7] \# | A[15]\# | REQ[1]\# | J |
| VSS | TRST\# | BPM[0]\# | PRDY\# | A[14]\# | ADSTB[0]\# | VSS | K |
| RSVD | TDO_M | TMS | VSS | VSS | A[12]\# | A[16]\# | L |
| TDI_M | TDO | TCK | RSVD | A[10]\# | A[13]\# | VSS | M |
| FORCEPR\# | TDI | VSS | SLP\# | A[8] \# | A[5] \# | REQ[0]\# | N |
| VSS | VSS | RSVD | VSS | VSS | REQ[3] \# | A[3] \# | P |
| LINT1 | STPCLK\# | DPSLP\# | DPRSTP\# | REQ[4]\# | A[6]\# | VSS | R |
| LINTO | FERR\# | RSVD | VSS | DRDY\# | BRO\# | DEFER\# | T |
| VSS | VSS | SMI\# | A20M | VSS | RS[2]\# | BPRI\# | U |
| BR1\# | INIT\# | PWRGOOD | VSS | ADS | HITM\# | VSS | v |
| D[10]\# | DINV[0] | VSS | RS[0]\# | TRDY\# | LOCK\# | VSS | w |
| DSTBP[0] | D[12]\# | RS[1]\# | DBSY\# | BNR\# | VSS | VSS | Y |
| VSS | D[6]\# | HIT\# | VSS | VSS | VSS |  | AA |
| 15 | 16 | 17 | 18 | 19 | 20 | 21 |  |

Table 4-12.Pinout Arranged By Signal Name (Sheet 1 of 4)

| Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A[10]\# | M19 | BCLK[0] | V11 | D[28]\# | T3 | D[62]\# | C7 |
| A[11]\# | H21 | BCLK[1] | V12 | D[29]\# | AA8 | D[63]\# | D2 |
| A[12]\# | L20 | BNR\# | Y19 | D[3]\# | AA14 | D[7]\# | Y10 |
| A[13]\# | M20 | BPM[0]\# | K17 | D[30]\# | V2 | D[8]\# | Y9 |
| A[14]\# | K19 | BPM[1]\# | J18 | D[31]\# | W4 | D[9]\# | Y13 |
| A[15]\# | J20 | BPM[2]\# | H15 | D[32]\# | R3 | DBSY\# | Y18 |
| A[16]\# | L21 | BPM[3] \# | J15 | D[33]\# | R2 | RSVD | V5 |
| A[17]\# | C19 | BPRI\# | U21 | D[34]\# | P1 | DEFER\# | T21 |
| A[18]\# | F19 | BRO\# | T20 | D[35]\# | N1 | DINV[0]\# | W16 |
| A[19]\# | E21 | BR1\# | V15 | D[36]\# | M2 | DINV[1]\# | Y6 |
| A[20]\# | A16 | BSEL[0] | J6 | D[37]\# | P2 | DINV[2]\# | L1 |
| A[21]\# | D19 | BSEL[1] | H5 | D[38]\# | J3 | DINV[3]\# | C5 |
| A[22]\# | C14 | BSEL[2] | G5 | D[39]\# | N3 | DPRSTP\# | R18 |
| A[23]\# | C18 | COMP[0] | T1 | D[4]\# | AA11 | DPWR\# | U4 |
| A[24]\# | C20 | COMP[1] | T2 | D[40]\# | G3 | DRDY\# | T19 |
| A[25]\# | E20 | COMP[2] | F20 | D[41]\# | H2 | DSTBN[0]\# | Y14 |
| A[26]\# | D20 | COMP[3] | F21 | D[42]\# | N2 | DSTBN[1]\# | Y4 |
| A[27]\# | B18 | D[0]\# | Y11 | D[43]\# | L2 | DSTBN[2]\# | K2 |
| A[28]\# | C15 | D[1]\# | W10 | D[44]\# | M3 | DSTBN[3]\# | E2 |
| A[29]\# | B16 | D[10]\# | W15 | D[45]\# | J2 | DSTBP[0]\# | Y15 |
| A[3]\# | P21 | D[11]\# | AA13 | D[46]\# | H1 | DSTBP[1]\# | Y5 |
| A[30]\# | B17 | D[12]\# | Y16 | D[47]\# | J1 | DSTBP[2]\# | K3 |
| A[31]\# | C16 | D[13]\# | W13 | D[48]\# | C2 | DSTBP[3]\# | F3 |
| A[32]\# | A17 | D[14]\# | AA9 | D[49]\# | G2 | FERR\# | T16 |
| A[33]\# | B14 | D[15]\# | W9 | D[5]\# | W12 | FORCEPR\# | N15 |
| A[34]\# | B15 | D[16]\# | AA5 | D[50]\# | F1 | GTLREF_MA | A7 |
| A[35]\# | A14 | D[17]\# | Y8 | D[51]\# | D3 | HIT\# | AA17 |
| A[4]\# | H20 | D[18]\# | W3 | D[52]\# | B4 | HITM\# | V20 |
| A[5]\# | N20 | D[19]\# | U1 | D[53]\# | E1 | IERR\# | F16 |
| A[6]\# | R20 | D[2]\# | Y12 | D[54]\# | A5 | IGNNE\# | J4 |
| A[7]\# | J19 | D[20]\# | W7 | D[55]\# | C3 | INIT\# | V16 |
| A[8]\# | N19 | D[21]\# | W6 | D[56]\# | A6 | LINTO | T15 |
| A[9]\# | G20 | D[22]\# | Y7 | D[57]\# | F2 | LINT1 | R15 |
| A20M\# | U18 | D[23]\# | AA6 | D[58]\# | C6 | LOCK\# | W20 |
| RSVD | U5 | D[24]\# | Y3 | D[59]\# | B6 | RSVD | P17 |
| ADS\# | V19 | D[25] \# | W2 | D[6]\# | AA16 | GTLREF_EA | D6 |
| ADSTB[0]\# | K20 | D[26]\# | V3 | D[60]\# | B3 | BPM_2[0]\# | G6 |

Table 4-12.Pinout Arranged By Signal Name (Sheet 2 of 4)

| Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADSTB[1]\# | B19 | D[27]\# | U2 | D[61]\# | C4 | BPM_2[1]\# | H6 |
| BPM_2[2]\# | K4 | TDO | M16 | VCC | L10 | VSS | B13 |
| BPM_2[3]\# | K5 | EXTBGREF | M6 | VCC | L11 | VSS | B20 |
| TDI_M | M15 | THERMTRIP\# | H17 | VCC | L12 | VSS | B21 |
| TDO_M | L16 | THRMDA | E4 | VCC | M10 | VSS | C8 |
| PRDY\# | K18 | THRMDC | E5 | VCC | M11 | VSS | C17 |
| PREQ\# | J16 | TMS | L17 | VCC | M12 | VSS | D1 |
| PROCHOT\# | G17 | TRDY\# | W19 | VCC | N10 | VSS | D5 |
| PWRGOOD | V17 | TRST\# | K16 | VCC | N11 | VSS | D8 |
| REQ[0]\# | N21 | VCC | A10 | VCC | N12 | VSS | D14 |
| REQ[1] \# | J21 | VCC | A11 | VCC | P10 | VSS | D18 |
| REQ[2] \# | G19 | VCC | A12 | VCC | P11 | VSS | D21 |
| REQ[3]\# | P20 | VCC | B10 | VCC | P12 | VSS | E3 |
| REQ[4]\# | R19 | VCC | B11 | VCC | R10 | VSS | E6 |
| RESET\# | D15 | VCC | B12 | VCC | R11 | VSS | E7 |
| RS[0]\# | W18 | VCC | C10 | VCC | R12 | VSS | E8 |
| RS[1]\# | Y17 | VCC | C11 | VCCA | D7 | VSS | E15 |
| RS[2]\# | U20 | VCC | C12 | VTT | V10 | VSS | E16 |
| RSVD | D17 | VCC | D10 | VCCQ0 | A9 | VSS | E19 |
| DPSLP\# | R17 | VCC | D11 | VCCQ0 | B9 | THRMDC_2 | F4 |
| RSVD | M18 | VCC | D12 | VCCSENSE | C13 | VSS | F5 |
| RSVD | T17 | VCC | E10 | VID[0] | F15 | VSS | F6 |
| CMREF | B7 | VCC | E11 | VID[1] | D16 | VSS | F7 |
| RSVD | A13 | VCC | E12 | VID[2] | E18 | VSS | F17 |
| RSVD | R6 | VCC | F10 | VID[3] | G15 | VSS | F18 |
| RSVD | N6 | VCC | F11 | VID[4] | G16 | VSS | G1 |
| RSVD | T6 | VCC | F12 | VID[5] | E17 | THRMDA_2 | G4 |
| RSVD | A3 | VCC | G10 | VID[6] | G18 | VSS | G7 |
| RSVD | C1 | VCC | G11 | VSS | A2 | VSS | G9 |
| RSVD | C21 | VCC | G12 | VSS | A4 | VSS | G13 |
| VTT | E13 | VCC | H10 | VSS | A8 | VSS | G21 |
| VTT | E14 | VCC | H11 | VSS | A15 | VSS | H3 |
| VTT | F13 | VCC | H12 | VSS | A18 | VSS | H4 |
| VTT | F14 | VCC | J10 | RSVD | A19 | VSS | H7 |
| SLP\# | N18 | VCC | J11 | VSS | A20 | VSS | H9 |
| SMI\# | U17 | VCC | J12 | VSS | B1 | VSS | H13 |
| STPCLK\# | R16 | VCC | K10 | VSS | B2 | VSS | H16 |

Table 4-12.Pinout Arranged By Signal Name (Sheet 3 of 4)

| Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TCK | M17 | VCC | K11 | VSS | B5 | VSS | H18 |
| TDI | N16 | VCC | K12 | VSS | B8 | VSS | H19 |
| VSS | J5 | VSS | P7 | VSS | W5 | VTT | N8 |
| VSS | J7 | VSS | P9 | VSS | W8 | VTT | N14 |
| VSS | J9 | VSS | P13 | VSS | W11 | VTT | P8 |
| VSS | J13 | VSS | P15 | VSS | W14 | VTT | P14 |
| VSS | J17 | VSS | P16 | VSS | W17 | VTT | R8 |
| VSS | K1 | VSS | P18 | VSS | W21 | VTT | R14 |
| VSS | K6 | VSS | P19 | VSS | Y1 | VTT | T8 |
| VSS | K7 | VSS | R1 | VSS | Y2 | VTT | T14 |
| VSS | K9 | VSS | R5 | VSS | Y20 | VTT | U8 |
| VSS | K13 | VSS | R7 | VSS | Y21 | VTT | U9 |
| VSS | K15 | VSS | R9 | VSS | AA2 | VTT | U10 |
| VSS | K21 | VSS | R13 | VSS | AA3 | VTT | U11 |
| VSS | L3 | VSS | R21 | VSS | AA4 | VTT | U12 |
| VSS | L4 | VSS | T4 | VSS | AA7 | VTT | U13 |
| COMP_2[0] | L5 | RSVD | T5 | VSS | AA10 | VTT | U14 |
| VSS | L6 | VSS | T7 | VSS | AA12 | RSVD | V9 |
| VSS | L7 | VSS | T9 | VSS | AA15 | RSVD | R4 |
| VSS | L9 | VSS | T10 | VSS | AA18 | RSVD | M4 |
| VSS | L13 | VSS | T11 | VSS | AA19 | RSVD | D4 |
| RSVD | L15 | VSS | T12 | VSS | AA20 |  |  |
| VSS | L18 | VSS | T13 | VSSSENSE | D13 |  |  |
| VSS | L19 | VSS | T18 | VTT | C9 |  |  |
| VSS | M1 | VSS | U3 | VTT | D9 |  |  |
| VSS | M5 | VSS | U6 | VTT | E9 |  |  |
| VSS | M7 | VSS | U7 | VTT | F8 |  |  |
| VSS | M9 | VSS | U15 | VTT | F9 |  |  |
| VSS | M13 | VSS | U16 | VTT | G8 |  |  |
| VSS | M21 | VSS | U19 | VTT | G14 |  |  |
| COMP_2[2] | N4 | VSS | V1 | VTT | H8 |  |  |
| COMP_2[1] | N5 | VSS | V4 | VTT | H14 |  |  |
| VSS | N7 | VSS | V6 | VTT | J8 |  |  |
| VSS | N9 | VSS | V7 | VTT | J14 |  |  |
| VSS | N13 | VSS | V8 | VTT | K8 |  |  |
| VSS | N17 | VSS | V13 | VTT | K14 |  |  |
| VSS | P3 | VSS | V14 | VTT | L8 |  |  |

Table 4-12.Pinout Arranged By Signal Name (Sheet 4 of 4)

| Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# | Signal Name | Ball \# |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| COMP_2[3] | P4 | VSS | V18 | VTT | L14 |  |  |
| RSVD | P5 | VSS | V21 | VTT | M8 |  |  |
| VSS | P6 | VSS | W1 | VTT | M14 |  |  |

### 4.3 Signal Description

Table 4-13.Signal Description (Sheet 1 of 8)

| Signal Name | Type | $\quad$ Description |
| :--- | :---: | :--- |
| A[32:3]\# | I/O | A[32:3]\# (Address) defines a 232-byte physical memory address <br> space. In subphase 1 of the address phase, these pins transmit the <br> address of a transaction. <br> In sub-phase 2, these pins transmit transaction type information. <br> These signals must connect the appropriate pins of both agents on <br> the processor FSB. A[32:3]\# are source synchronous signals and <br> are latched into the receiving buffers by ADSTB[1:0]\#. Address <br> signals are used as straps which are sampled before RESET\# is <br> deasserted. |
| A20M\# | I | If A20M\# (Address-20 Mask) is asserted, the processor masks <br> physical address bit 20 (A20\#) before looking up a line in any <br> internal cache and before driving a read/write transaction on the <br> bus. Asserting A20M\# emulates the 8086 processor's address <br> wrap-around at the 1-MB boundary. Assertion of A20M\# is only <br> supported in real mode. <br> A20M\# is an asynchronous signal. However, to ensure recognition <br> of this signal following an input/output write instruction, it must be <br> valid along with the TRDY\# assertion of the corresponding input/ <br> output Write bus transaction. |
| ADS\# | I/O | ADS\# (Address Strobe) is asserted to indicate the validity of the <br> transaction address on the A[32:3]\# and REQ[4:0]\# pins. All bus <br> agents observe the ADS\# activation to begin parity checking, <br> protocol checking, address decode, internal loop, or deferred reply <br> ID match operations associated with the new transaction. |
| ADSTB[1:0]\# | I/O | Address strobes are used to latch A[32:3]\# and REQ[4:0]\# on <br> their rising and falling edges. Strobes are associated with signals as <br> shown below. <br> SignalsAssociated Strobe <br> REQ[4:0]\#, A[16:3]\#ADSTB[0]\# <br> A[32:17]\#ADSTB[1]\# |
| BNR\#\#[1:0] | I | The differential pair BCLK (Bus Clock) determines the FSB <br> frequency. All FSB agents must receive these signals to drive their <br> outputs and latch their inputs. All external timing parameters are <br> specified with respect to the rising edge of BCLK0 crossing <br> VCROSS. |
| I/O | BNR\# (Block Next Request) is used to assert a bus stall by any bus <br> agent who is unable to accept new bus transactions. During a bus <br> stall, the current bus owner cannot issue any new transactions. |  |

Table 4-13.Signal Description (Sheet 2 of 8)

| Signal Name | Type | $\quad$ Description |
| :--- | :---: | :--- |$|$| BPM[0]\#, <br> BPM_2[0]\# | O | BPM[3:0]\# (Breakpoint Monitor) are breakpoint and performance <br> monitor signals. They are outputs from the processor which <br> indicate the status of breakpoints and programmable counters used <br> for monitoring processor performance. BPM[3:0]\# should connect <br> the appropriate pins of all FSB agents.This includes debug or <br> performance monitoring tools. Refer to the platform design guide <br> for more detailed information. <br> * BPM_2[3:0] refer to core \#2. |
| :--- | :---: | :--- |
| BPM[1]\#, <br> BPM_2[1]\#, | I/O | BPM[2]\#, <br> BPM_2[2]\# |
| BPM[3]\#, <br> BPM_2[3]\# | I/O | I <br> BPRI\# |
| BRRI\# (Bus Priority Request) is used to arbitrate for ownership of |  |  |
| the FSB. It must connect the appropriate pins of both FSB agents. |  |  |
| Observing BPRI\# active (as asserted by the priority agent) causes |  |  |
| the other agent to stop issuing new requests, unless such requests |  |  |
| are part of an ongoing locked operation. The priority agent keeps |  |  |
| BPRI\# asserted until all of its requests are completed, then |  |  |
| releases the bus by deasserting BPRI\#. |  |  |

Package Mechanical Specifications and Ball Information

Table 4-13.Signal Description (Sheet 3 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| DEFER\# | I | DEFER\# is asserted by an agent to indicate that a transaction cannot be guaranteed in-order completion. Assertion of DEFER\# is normally the responsibility of the addressed memory or Input/ Output agent. This signal must connect the appropriate pins of both FSB agents. |
| DINV[3:0]\# | I | DINV[3:0]\# (Data Bus Inversion) are source synchronous and indicate the polarity of the D[63:0]\# signals. The DINV[3:0]\# signals are activated when the data on the data bus is inverted. The bus agent will invert the data bus signals if more than half the bits, within the covered group, would change level in the next cycle. <br> DINV[3:0]\# Assignment To Data Bus <br> Bus SignalData Bus Signals <br> DINV[3]\#D[63:48]\# <br> DINV[2]\#D[47:32]\# <br> DINV[1]\#D[31:16]\# <br> DINV[0]\# D[15:0]\# |
| DPSLP\# | I | DPSLP\# when asserted on the platform causes the processor to transition from the Sleep State to the Deep Sleep state. In order to return to the Sleep State, DPSLP\# must be deasserted. DPSLP\# is driven by the SCH chipset. This signal is not used for Entry Level Desktop platform '08 and tied to VTT. |
| DPRSTP\# | I | DPRSTP\# when asserted on the platform causes the processor to transition from the Deep Sleep State to the Deeper Sleep state. In order to return to the Deep Sleep State, DPRSTP\# must be deasserted. DPRSTP\# is driven by the SCH chipset. This signal is not used for Entry Level Desktop platform '08 and tied to VTT through a 1 k Ohm resistor. |
| DPWR\# | I | DPWR\# is a control signal from the chipset used to reduce power on the processor data bus input buffers. This signal is not used for Entry Level Desktop platform '08 and tied to VTT. |
| DRDY\# | I/O | DRDY\# (Data Ready) is asserted by the data driver on each data transfer, indicating valid data on the data bus. In a multi-common clock data transfer, DRDY\# may be deasserted to insert idle clocks. This signal must connect the appropriate pins of both FSB agents. |
| DSTBN[3:0]\# | I/O | Data strobe used to latch in D[63:0]\#. <br> SignalsAssociated Strobe <br> D[15:0]\#DINV[0]\#, DSTBN[0]\# <br> D[31:16]\#DINV[1]\#, DSTBN[1]\# <br> D[47:32]\#DINV[2]\#, DSTBN[2]\# <br> D[63:48]\#DINV[3]\#, DSTBN[3]\# |
| DSTBP[3:0]\# | I/O | Data strobe used to latch in D[63:0]\#. <br> SignalsAssociated Strobe <br> D[15:0]\#DINV[0]\#, DSTBP[0]\# <br> D[31:16]\#DINV[1]\#, DSTBP[1]\# <br> D[47:32]\#DINV[2]\#, DSTBP[2]\# <br> D[63:48]\#DINV[3]\#, DSTBP[3]\# |

Package Mechanical Specifications and Ball Information

Table 4-13.Signal Description (Sheet 4 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| FERR\#/PBE\# | 0 | FERR\# (Floating-point Error)/PBE\# (Pending Break Event) is a multiplexed signal and its meaning is qualified with STPCLK\#. When STPCLK\# is not asserted, FERR\#/PBE\# indicates a floating point when the processor detects an unmasked floating-point error. FERR\# is similar to the ERROR\# signal on the Intel 387 coprocessor, and is included for compatibility with systems using MSDOS*- type floating-point error reporting. When STPCLK\# is asserted, an assertion of FERR\#/PBE\# indicates that the processor has a pending break event waiting for service. The assertion of FERR\#/PBE\# indicates that the processor should be returned to the Normal state. When FERR\#/PBE\# is asserted, indicating a break event, it will remain asserted until STPCLK\# is deasserted. Assertion of PREQ\# when STPCLK\# is active will also cause an FERR\# break event. <br> For additional information on the pending break event functionality, including identification of support of the feature and enable/disable information, refer to Volume 3 of the Intel ${ }^{\circledR} 64$ and IA-32 Architectures Software Developer's Manuals and the Intel ${ }^{\circledR}$ Processor Identification and CPUID Instruction Application Note. For termination requirements, refer to the platform design guide. |
| CMREF | PWR | CMREF should be set the same as GTLREF_*. |
| EXTBGREF | PWR | EXTBGREF should be set at $2 / 3 \mathrm{~V}_{\text {Tr }}$. |
| GTLREF_MA | PWR | GTLREF Middle Agent. Refer to the platform design guide for details on GTLREF implementation. GTLREF_MA should have separate voltage divider resistor network to set the right reference voltage and cannot share the voltage divider resistor network with GTLREF_EA. |
| GTLREF_EA | PWR | GTLREF End Agent. Refer to the platform design guide for details on GTLREF implementation. GTLREF_EA should have separate voltage divider resistor network to set the right reference voltage and cannot share the voltage divider resistor network with GTLREF_MA. |
| HIT\# <br> HITM\# | I/O | HIT\# (Snoop Hit) and HITM\# (Hit Modified) convey transaction snoop operation results. Either FSB agent may assert both HIT\# and HITM\# together to indicate that it requires a snoop stall, which can be continued by reasserting HIT\# and HITM\# together. |
| IERR\# | 0 | IERR\# (Internal Error) is asserted by a processor as the result of an internal error. Assertion of IERR\# is usually accompanied by a SHUTDOWN transaction on the FSB. This transaction may optionally be converted to an external error signal (e.g., NMI) by system core logic. The processor will keep IERR\# asserted until the assertion of RESET\#, or INIT\#. <br> For termination requirements, refer to the platform design guide. |

Table 4-13.Signal Description (Sheet 5 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| IGNNE\# | I | IGNNE\# (Ignore Numeric Error) is asserted to force the processor to ignore a numeric error and continue to execute non-control floating-point instructions. If IGNNE\# is deasserted, the processor generates an exception on a non-control floating-point instruction if a previous floating-point instruction caused an error. IGNNE\# has no effect when the NE bit in control register 0 (CRO) is set. <br> IGNNE\# is an asynchronous signal. However, to ensure recognition of this signal following an Input/Output write instruction, it must be valid along with the TRDY\# assertion of the corresponding Input/ Output Write bus transaction. |
| INIT\# | I | INIT\# (Initialization), when asserted, resets integer registers inside the processor without affecting its internal caches or floating-point registers. The processor then begins execution at the power-on Reset vector configured during power-on configuration. The processor continues to handle snoop requests during INIT\# assertion. INIT\# is an asynchronous signal. However, to ensure recognition of this signal following an Input/Output Write instruction, it must be valid along with the TRDY\# assertion of the corresponding Input/Output Write bus transaction. INIT\# must connect the appropriate pins of both FSB agents. <br> If INIT\# is sampled active on the active to inactive transition of RESET\#, processor reverses its FSB data and address signals internally to ease mother board layout for systems where the chipset is on the other side of the mother board. <br> $D[63: 0]=>D[0: 63]$ <br> A[32:3] => A[3:32] <br> DINV[3:0]\# is also reversed. |
| LINT[1:0] | I | LINT[1:0] (Local APIC Interrupt) must connect the appropriate pins of all APIC Bus agents. When the APIC is disabled, the LINT0 signal becomes INTR, a maskable interrupt request signal, and LINT1 becomes NMI, a non-maskable interrupt. INTR and NMI are backward compatible with the signals of those names on the Pentium processor. Both signals are asynchronous. <br> Both of these signals must be software configured via BIOS programming of the APIC register space to be used either as NMI/ INTR or LINT[1:0]. Because the APIC is enabled by default after Reset, operation of these pins as LINT[1:0] is the default configuration. |
| LOCK\# | I/O | Probe Ready signal used by debug tools to determine processor debug readiness. |
| PRDY\# | 0 | Probe Request signal used by debug tools to request debug operation of the processor. Refer to the platform design guide for more implementation details. |
| PREQ\# | I | Probe Request signal used by debug tools to request debug operation of the processor. Refer to the platform design guide for more implementation details. |

Package Mechanical Specifications and Ball Information

Table 4-13.Signal Description (Sheet 6 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| PROCHOT\# | $\begin{gathered} \mathrm{I} / \mathrm{O}, \mathrm{O} \\ \text { (DP) } \end{gathered}$ | As an output, PROCHOT\# (Processor Hot) will go active when the processor temperature monitoring sensor detects that the processor has reached its maximum safe operating temperature. This indicates that the processor Thermal Control Circuit (TCC) has been activated, if enabled. As an input, assertion of PROCHOT\# by the system will activate the TCC, if enabled. The TCC will remain active until the system deasserts PROCHOT\#. <br> For termination requirements, refer to the platform design guide. This signal may require voltage translation on the motherboard. Refer to the platform design guide for more details. |
| PWRGOOD | I | PWRGOOD (Power Good) is a processor input. The processor requires this signal to be a clean indication that the clocks and power supplies are stable and within their specifications. 'Clean' implies that the signal will remain low (capable of sinking leakage current), without glitches, from the time that the power supplies are turned on until they come within specification. The signal must then transition monotonically to a high state. PWRGOOD can be driven inactive at any time, but clocks and power must again be stable before a subsequent rising edge of PWRGOOD. <br> The PWRGOOD signal must be supplied to the processor; it is used to protect internal circuits against voltage sequencing issues. It should be driven high throughout boundary scan operation. |
| REQ[4:0]\# | I/O | REQ[4:0]\# (Request Command) must connect the appropriate pins of both FSB agents. They are asserted by the current bus owner to define the currently active transaction type. These signals are source synchronous to ADSTB[0]\#. |
| RESET\# | I | Asserting the RESET\# signal resets the processor to a known state and invalidates its internal caches without writing back any of their contents. For a power-on Reset, RESET\# must stay active for at least two milliseconds after $\mathrm{V}_{\mathrm{CC}}$ and BCLK have reached their proper specifications. On observing active RESET\#, both FSB agents will deassert their outputs within two clocks. All processor straps must be valid within the specified setup time before RESET\# is deasserted. |
| RS[2:0]\# | I | RS[2:0]\# (Response Status) are driven by the response agent (the agent responsible for completion of the current transaction), and must connect the appropriate pins of both FSB agents. |
| RSVD | Reserved/ No Connect | These pins are RESERVED and must be left unconnected on the board. However, it is recommended that routing channels to these pins on the board be kept open for possible future use. |

Table 4-13.Signal Description (Sheet 7 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| SLP\# | I | SLP\# (Sleep), when asserted in Stop-Grant state, causes the processor to enter the Sleep state. During Sleep state, the processor stops providing internal clock signals to all units, leaving only the Phase-Locked Loop (PLL) still operating. Processors in this state will not recognize snoops or interrupts. The processor will recognize only assertion of the RESET\# signal, deassertion of SLP\#, and removal of the BCLK input while in Sleep state. If SLP\# is deasserted, the processor exits Sleep state and returns to StopGrant state, restarting its internal clock signals to the bus and processor core units. If DPSLP\# is asserted while in the Sleep state, the processor will exit the Sleep state and transition to the Deep Sleep state. This signal is not used for Entry Level Desktop platform '08 and tied to VTT. |
| SMI\# | I | SMI\# (System Management Interrupt) is asserted asynchronously by system logic. On accepting a System Management Interrupt, the processor saves the current state and enter System Management Mode (SMM). An SMI Acknowledge transaction is issued, and the processor begins program execution from the SMM handler. If SMI\# is asserted during the deassertion of RESET\# the processor will tristate its outputs. |
| STPCLK\# | I | RSVD. |
| TCK | I | TCK (Test Clock) provides the clock input for the processor Test Bus (also known as the Test Access Port). Refer to the platform design guide for termination requirements and implementation details. |
| TDI, TDI_M | I | TDI (Test Data In) transfers serial test data into the processor. TDI provides the serial input needed for JTAG specification support. <br> * TDI_M refers to core \#2. |
| TDO, TDO_M | 0 | TDO (Test Data Out) transfers serial test data out of the processor. TDO provides the serial output needed for JTAG specification support. <br> * TDO_M refers to core \#2. |
| THRMTRIP\# | 0 | The processor protects itself from catastrophic overheating by use of an internal thermal sensor. This sensor is set well above the normal operating temperature to ensure that there are no false trips. The processor will stop all execution when the junction temperature exceeds approximately $125^{\circ} \mathrm{C}$. This is signalled to the system by the THERMTRIP\# (Thermal Trip) pin. For termination requirements, refer to the platform design guide. |
| THRMDA, THRMDA_2 | PWR | Thermal Diode - Anode <br> * THRMDA_2 refers to core \#2. |
| THRMDC, THRMDC_2 | PWR | Thermal Diode - Cathode <br> * THRMDC_2 refers to core \#2. |
| TMS | I | TMS (Test Mode Select) is a JTAG specification support signal used by debug tools. Refer to the platform design guide for termination requirements and implementation details. |
| TRDY\# | I | TRDY\# (Target Ready) is asserted by the target to indicate that it is ready to receive a write or implicit writeback data transfer. TRDY\# must connect the appropriate pins of both FSB agents. |

Package Mechanical Specifications and Ball Information

Table 4-13.Signal Description (Sheet 8 of 8)

| Signal Name | Type | Description |
| :---: | :---: | :---: |
| TRST\# | I | TRST\# (Test Reset) resets the Test Access Port (TAP) logic. TRST\# must be driven low during power on Reset. Refer to the platform design guide for termination requirements and implementation details. |
| $\mathrm{V}_{\text {CCA }}$ | PWR | $\mathrm{V}_{\text {CCA }}$ provides isolated power for the internal processor core PLLs. Refer to the platform design guide for complete implementation details. |
| $\mathrm{V}_{\text {CC }}$ | PWR | Processor core power supply |
| $\mathrm{V}_{\text {SS }}$ | GND | Processor core ground node. |
| VID[6:0] | 0 | VID[6:0] (Voltage ID) pins are used to support automatic selection of power supply voltages $\left(\mathrm{V}_{\mathrm{CC}}\right)$ but these pins are not used in the Entry Level Desktop platform '08 as the VID is fixed at 1.1 V . |
| $\mathrm{V}_{\text {T }}$ | PWR | AGTL+ reference voltage |
| $\mathrm{V}_{\text {CC_SENSE }}$ | 0 | VCC_SENSE is an isolated low impedance connection to processor core power ( $\mathrm{V}_{\mathrm{CCP}}$ ). It can be used to sense or measure voltage near the silicon with little noise. |
| $\mathrm{V}_{\text {SS_SENSE }}$ | 0 | $\mathrm{V}_{\text {SS_SENSE }}$ is an isolated low impedance connection to processor core $\mathrm{V}_{\mathrm{SS}}$. It can be used to sense or measure ground near the silicon with little noise. Refer to the platform design guide for termination recommendations and more details. |
| vCCQO, <br> VCCQ1 | PWR | Connect this to $\mathrm{V}_{\text {T }}$. |

## 5 Thermal Specifications and Design Considerations

The processor requires a thermal solution to maintain temperatures within operating limits as set forth in Section Thermal Specifications. Any attempt to operate the processor outside these operating limits may result in permanent damage to the processor and potentially other components in the system. As processor technology changes, thermal management becomes increasingly crucial when building computer systems. Maintaining the proper thermal environment is key to reliable, long-term system operation.

A complete thermal solution includes both component and system level thermal management features. Component level thermal solutions include active or passive heatsink attached to the exposed processor die. The solution should make firm contact to the die while maintaining processor mechanical specifications such as pressure. A typical system level thermal solution may consist of a system fan used to evacuate or pull air through the system. For more information on designing a component level thermal solution, please refer to the appropriate Thermal and Mechanical Design Guidelines (see Section 1.3). Alternatively, the processor may be in a fan-less system, but would likely still use a multi-component heat spreader. Note that trading of thermal solutions also involves trading performance.

### 5.1 Thermal Specifications

To allow for the optimal operation and long-term reliability of Intel processor based systems, the system/processor thermal solution should be designed such that the processor remains within minimum and maximum case temperature (Tc) specification at the corresponding thermal design power (TDP) value listed in Table 5-14. Thermal solutions not designed to provide this level of thermal capability may affect the longterm reliability of the processor and system. For more details on thermal solution design, refer to the appropriate Thermal and Mechanical Design Guidelines (see Section 1.3).

The case temperature is defined at the geometric top center of the processor. Analysis indicates that real applications are unlikely to cause the processor to consume the theoretical maximum power dissipation for sustained time periods. Intel recommends that complete thermal solution designs target the TDP indicated in Table 5-14 instead of the maximum processor power consumption. The Intel Thermal Monitor feature is designed to help protect the processor in the unlikely event that an application exceeds the TDP recommendation for a sustained period of time. For more details on the usage of this feature, refer to Section 5.1.2. In all cases, the Intel Thermal Monitor feature must be enabled for the processor to remain within specification.

Thermal Specifications and Design Considerations

Table 5-14. Power Specifications for the Standard Voltage processor

| Symbol | Processor <br> Number | Core <br> Frequency <br> and Voltage | Thermal Design <br> Power |  | Unit | Tc <br> min <br> $\left({ }^{\circ} \mathrm{C}\right)$ | Tc <br> max <br> $\left({ }^{\circ} \mathrm{C}\right)$ | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TDP | 330 | $1.6 \mathrm{GHz} \& \mathrm{~V}_{\mathrm{CC}}$ | 8.0 |  | W | 0 | 85.2 | $1,3,4$ |
| Symbol | Parameter |  | Min | Typ | Max | Unit |  |  |
| PAH | Auto Halt | - | - | 2 | W |  |  | 2 |

NOTES:

1. The TDP specification should be used to design the processor thermal solution. The TDP is not the maximum theoretical power the processor can generate.
2. Not $100 \%$ tested. These power specifications are determined by characterization of the processor currents at higher temperatures and extrapolating the values for the temperature indicated.
3. The Intel Thermal Monitor automatic mode must be enabled for the processor to operate within specifications.
4. $\quad \mathrm{V}_{\mathrm{CC}}$ is determined by processor VID[6:0].

### 5.1.1 Thermal Diode

The processor incorporates an on-die PNP transistor whose base emitter junction is used as a thermal "diode", with its collector shorted to ground. The thermal diode can be read by an off-die analog/digital converter (a thermal sensor) located on the motherboard or a stand-alone measurement kit. The thermal diode may be used to monitor the die temperature of the processor for thermal management or instrumentation purposes but is not a reliable indication that the maximum operating temperature of the processor has been reached. When using the thermal diode, a temperature offset value must be read from a processor MSR and applied. See Section 5.1.2 for more details. See Section Section 5.1.2 for thermal diode usage recommendation when the PROCHOT\# signal is not asserted.

The reading of the external thermal sensor (on the motherboard) connected to the processor thermal diode signals will not necessarily reflect the temperature of the hottest location on the die. This is due to inaccuracies in the external thermal sensor, on-die temperature gradients between the location of the thermal diode and the hottest location on the die, and time based variations in the die temperature measurement. Time based variations can occur when the sampling rate of the thermal diode (by the thermal sensor) is slower than the rate at which the $\mathrm{T}_{\mathrm{J}}$ temperature can change.

Offset between the thermal diode based temperature reading and the Intel Thermal Monitor reading may be characterized using the Intel Thermal Monitor's Automatic mode activation of the thermal control circuit. This temperature offset must be taken into account when using the processor thermal diode to implement power management events. This offset is different than the diode $T_{\text {offset }}$ value programmed into the processor Model Specific Register (MSR).

Table 5-15 and Table 5-16 provide the diode interface and specifications. Transistor model parameters shown in Table 5-16 providing more accurate temperature measurements when the diode ideality factor is closer to the maximum or minimum limits. Contact your external sensor supplier for their recommendation. The thermal diode is separate from the Thermal Monitor's thermal sensor and cannot be used to predict the behavior of the Thermal Monitor.

Table 5-15.Thermal Diode Interface

| Signal Name | Pin/Ball Number | Signal Description |
| :---: | :---: | :---: |
| THERMDA | E4 | Thermal diode anode |
| THERMDC | E5 | Thermal diode cathode |

Table 5-16.Thermal Diode Parameters using Transistor Model

| Symbol | Parameter | Min | Typ | Max | Unit | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| IFW | Forward Bias Current | 5 |  | 200 | $\mu \mathrm{~A}$ | 1 |
| IE | Emitter Current | 5 |  | 200 | $\mu \mathrm{~A}$ | 1 |
| nQ | Transistor Ideality | 0.997 | 1.001 | 1.015 |  | $2,3,4$ |
| Beta |  | 0.25 |  | 0.65 |  | 2,3 |
| RT | Series Resistance | 2.79 | 4.52 | 6.24 | $\Omega$ | 2,5 |

## NOTES:

1. Intel does not support or recommend operation of the thermal diode under reverse bias.
2. Characterized across a temperature range of $50-100^{\circ} \mathrm{C}$.
3. Not $100 \%$ tested. Specified by design characterization.
4. The ideality factor, nQ , represents the deviation from ideal transistor model behavior as exemplified by the equation for the collector current:

$$
I_{C}=I_{S} *\left(e^{q V_{B E / ~} \mathrm{n}_{\mathrm{Q}} k T}-1\right)
$$

where $\mathrm{I}_{\mathrm{S}}=$ saturation current, $\mathrm{q}=$ electronic charge, $\mathrm{V}_{\mathrm{BE}}=$ voltage across the transistor base emitter junction (same nodes as VD), $\mathrm{k}=$ Boltzmann Constant, and $\mathrm{T}=$ absolute temperature (Kelvin).
5. The series resistance, $\mathrm{R}_{\mathrm{T}}$, provided in the Diode Model Table (Table 5-16) can be used for more accurate readings as needed.

When calculating a temperature based on the thermal diode measurements, a number of parameters must be either measured or assumed. Most devices measure the diode ideality and assume a series resistance and ideality trim value, although are capable of also measuring the series resistance. Calculating the temperature is then accomplished using the equations listed under Table 5-16. In most sensing devices, an expected value for the diode ideality is designed-in to the temperature calculation equation. If the designer of the temperature sensing device assumes a perfect diode, the ideality value (also called $n_{\text {trim }}$ ) will be 1.000 .

Given that most diodes are not perfect, the designers usually select an $n_{\text {trim }}$ value that more closely matches the behavior of the diodes in the processor. If the processor diode ideality deviates from that of the $n_{\text {trim }}$, each calculated temperature will be offset by a fixed amount. This temperature offset can be calculated with the equation:

$$
T_{\text {error }(n f)}=T_{\text {measured }} *\left(1-n_{\text {actual }} / n_{\text {trim }}\right)
$$

where $T_{\text {error(nf) }}$ is the offset in degrees $C, T_{\text {measured }}$ is in Kelvin, $n_{\text {actual }}$ is the measured ideality of the diode, and $n_{\text {trim }}$ is the diode ideality assumed by the temperature sensing device.

### 5.1.2 Intel ${ }^{\circledR}$ Thermal Monitor

The Intel Thermal Monitor helps control the processor temperature by activating the TCC (Thermal Control Circuit) when the processor silicon reaches its maximum operating temperature. The temperature at which the Intel Thermal Monitor activates the TCC is not user configurable. Bus traffic is snooped in the normal manner and interrupt requests are latched (and serviced during the time that the clocks are on) while the TCC is active.

With a properly designed and characterized thermal solution, it is anticipated that the TCC would only be activated for very short periods of time when running the most power intensive applications. The processor performance impact due to these brief periods of TCC activation is expected to be minor and hence not detectable. An underdesigned thermal solution that is not able to prevent excessive activation of the TCC in the anticipated ambient environment may cause a noticeable performance loss and may affect the long-term reliability of the processor. In addition, a thermal solution that is significantly under designed may not be capable of cooling the processor even when the TCC is active continuously.

The Intel Thermal Monitor controls the processor temperature by modulating (starting and stopping) the processor core clocks when the processor silicon reaches its maximum operating temperature. The Intel Thermal Monitor uses two modes to activate the TCC: automatic mode and on-demand mode. If both modes are activated, automatic mode takes precedence.

There is only one automatic modes called Intel Thermal Monitor 1 (TM1). This mode is selected by writing values to the MSRs of the processor. After automatic mode is enabled, the TCC will activate only when the internal die temperature reaches the maximum allowed value for operation.

The Intel Thermal Monitor automatic mode must be enabled through BIOS for the processor to be operating within specifications. Intel recommends TM1 be enabled on the processors.

When TM1 is enabled and a high temperature situation exists, the clocks will be modulated by alternately turning the clocks off and on at a $50 \%$ duty cycle. Cycle times are processor speed dependent and will decrease linearly as processor core frequencies increase. Once the temperature has returned to a non-critical level, modulation ceases and TCC goes inactive. A small amount of hysteresis has been included to prevent rapid
active/inactive transitions of the TCC when the processor temperature is near the trip point. The duty cycle is factory configured and cannot be modified. Also, automatic mode does not require any additional hardware, software drivers, or interrupt handling routines. Processor performance will be decreased by the same amount as the duty cycle when the TCC is active.

The Intel Thermal Monitor automatic mode must be enabled through BIOS for the processor to be operating within specifications. Intel recommends TM1 be enabled on the processors.

## TM1 feature is referred to as Adaptive Thermal Monitoring features.

The TCC may also be activated via on-demand mode. If bit 4 of the ACPI Intel Thermal Monitor control register is written to a 1 , the TCC will be activated immediately independent of the processor temperature. When using on-demand mode to activate the TCC, the duty cycle of the clock modulation is programmable via bits 3:1 of the same ACPI Intel Thermal Monitor control register. In automatic mode, the duty cycle is fixed at $50 \%$ on, $50 \%$ off, however in on-demand mode, the duty cycle can be programmed from $12.5 \%$ on/ $87.5 \%$ off, to $87.5 \%$ on/12.5\% off in $12.5 \%$ increments. On-demand mode may be used at the same time automatic mode is enabled, however, if the system tries to enable the TCC via on-demand mode at the same time automatic mode is enabled and a high temperature condition exists, automatic mode will take precedence.

An external signal, PROCHOT\# (processor hot) is asserted when the processor detects that its temperature is above the thermal trip point. Bus snooping and interrupt latching are also active while the TCC is active.

Besides the thermal sensor and thermal control circuit, the Intel Thermal Monitor also includes one ACPI register, one performance counter register, three MSR, and one I/O pin (PROCHOT\#). All are available to monitor and control the state of the Intel Thermal Monitor feature. The Intel Thermal Monitor can be configured to generate an interrupt upon the assertion or deassertion of PROCHOT\#.

PROCHOT\# will not be asserted when the processor is in the Stop Grant power states; hence, the thermal diode reading must be used as a safeguard to maintain the processor junction temperature within maximum specification. If the platform thermal solution is not able to maintain the processor junction temperature within the maximum specification, the system must initiate an orderly shutdown to prevent damage. If the processor enters one of the above power states with PROCHOT\# already asserted, PROCHOT\# will remain asserted until the processor exits the Stop Grant power state and the processor junction temperature drops below the thermal trip point.

If Intel Thermal Monitor automatic mode is disabled, the processor will be operating out of specification. Regardless of enabling the automatic or on-demand modes, in the event of a catastrophic cooling failure, the processor will automatically shut down when the silicon has reached a temperature of approximately $125^{\circ} \mathrm{C}$. At this point the THERMTRIP\# signal will go active. THERMTRIP\# activation is independent of processor activity and does not generate any bus cycles.

Thermal Specifications and Design Considerations

### 5.1.3 Digital Thermal Sensor

The processor also contains an on die Digital Thermal Sensor (DTS) that can be read via an MSR (no I/O interface). Each core of the processor will have a unique digital thermal sensor whose temperature is accessible via the processor MSRs. The DTS is the preferred method of reading the processor die temperature since it can be located much closer to the hottest portions of the die and can thus more accurately track the die temperature and potential activation of processor core clock modulation via the Thermal Monitor. The DTS is only valid while the processor is in the normal operating state (the Normal package level low power state).

Unlike traditional thermal devices, the DTS will output a temperature relative to the maximum supported operating temperature of the processor ( $\mathrm{T}_{\mathrm{J}} \max$ ). It is the responsibility of software to convert the relative temperature to an absolute temperature. The temperature returned by the DTS will always be at or below $T_{J_{\text {_ max }}}$. Catastrophic temperature conditions are detectable via an Out Of Spec status bit. This bit is also part of the DTS MSR. When this bit is set, the processor is operating out of specification and immediate shutdown of the system should occur. The processor operation and code execution is not ensured once the activation of the Out of Spec status bit is set.

The DTS-relative temperature readout corresponds to the Thermal Monitor (TM1) trigger point. When the DTS indicates maximum processor core temperature has been reached, the TM1 hardware thermal control mechanism will activate. The DTS and TM1temperature may not correspond to the thermal diode reading since the thermal diode is located in a separate portion of the die and thermal gradient between the individual core DTS. Additionally, the thermal gradient from DTS to thermal diode can vary substantially due to changes in processor power, mechanical and thermal attach, and software application. The system designer is required to use the DTS to ensure proper operation of the processor within its temperature operating specifications.

Changes to the temperature can be detected via two programmable thresholds located in the processor MSRs. These thresholds have the capability of generating interrupts via the core's local APIC. Refer to the Intel ${ }^{\circledR} 64$ and IA-32 Architectures Software Developer's Manuals for specific register and programming details.

### 5.1.4 Out of Specification Detection

Overheat detection is performed by monitoring the processor temperature and temperature gradient. This feature is intended for graceful shut down before the THERMTRIP\# is activated. If the processor's TM1 are triggered and the temperature remains high, an "Out Of Spec" status and sticky bit are latched in the status MSR register and generates thermal interrupt.

### 5.1.5 PROCHOT\# Signal Pin

An external signal, PROCHOT\# (processor hot), is asserted when the processor die temperature has reached its maximum operating temperature. If TM1 is enabled, then the TCC will be active when PROCHOT\# is asserted. The processor can be configured to
generate an interrupt upon the assertion or deassertion of PROCHOT\#. Refer to the Intel ${ }^{\circledR} 64$ and IA-32 Architectures Software Developer's Manuals for specific register and programming details.

The processor implements a bi-directional PROCHOT\# capability to allow system designs to protect various components from overheating situations. The PROCHOT\# signal is bi-directional in that it can either signal when the processor has reached its maximum operating temperature or be driven from an external source to activate the TCC. The ability to activate the TCC via PROCHOT\# can provide a means for thermal protection of system components.

Only a single PROCHOT\# pin exists at a package level of the processor. When the core's thermal sensor trips, PROCHOT\# signal will be driven by the processor package. If TM1 is enabled, PROCHOT\# will be asserted.It is important to note that Intel recommends TM1 to be enabled.

When PROCHOT\# is driven by an external agent, if TM1 is enabled on the core, then the processor core will have the clocks modulated.

PROCHOT\# may be used for thermal protection of voltage regulators (VR). System designers can create a circuit to monitor the VR temperature and activate the TCC when the temperature limit of the VR is reached. By asserting PROCHOT\# (pulled-low) and activating the TCC, the VR will cool down as a result of reduced processor power consumption. Bi-directional PROCHOT\# can allow VR thermal designs to target maximum sustained current instead of maximum current. Systems should still provide proper cooling for the VR and rely on bi-directional PROCHOT\# only as a backup in case of system cooling failure. The system thermal design should allow the power delivery circuitry to operate within its temperature specification even while the processor is operating at its TDP. With a properly designed and characterized thermal solution, it is anticipated that bi-directional PROCHOT\# would only be asserted for very short periods of time when running the most power intensive applications. An under-designed thermal solution that is not able to prevent excessive assertion of PROCHOT\# in the anticipated ambient environment may cause a noticeable performance loss.

Refer to the Voltage Regulation Specification for details on implementing the bidirectional PROCHOT\# feature.

## 6 Debug Tools Specifications

The ITP-XDP debug port connector is the recommended debug port for platforms using the processor. Contact your Intel representative for more information.

